

# Compact Broadband CPW-fed Taper-shaped Monopole Antenna with L-slots for C-band Applications

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**Abstract** A design approach for the compact broadband coplanar waveguide (CPW)-fed tapered monopole antenna (TMA) for C-band applications is presented, in this paper. The proposed antenna is composed of a CPW-fed monopole antenna with tapered structures, embedded with symmetric open-ended L-slots, and a tapered shape is cut from the top sides of the CPW-fed ground plane. By loading these structures, the return loss ( $S_{11}$ ) bandwidth is noticeably increased due to adjacent three resonance modes are generated from the monopole antenna. The designed TMA is simulated by electromagnetic simulator (CST Microwave Studio), with obtained optimal parameters, is demonstrated that  $S_{11} < -10$  dB impedance bandwidth is ranging from 3.8 to 8 GHz. The overall dimension of the antenna comes at  $20 \times 18 \times 1.6 \text{ mm}^3$ . This antenna has the advantages of simple configuration, low profile, compactness, and low fabrication cost. The design process and the parametric study of key parameters (taper and slot structures) in producing broadband antenna bandwidth are analyzed and discussed in detail. The simulated results demonstrate that the proposed TMA has good stable omnidirectional radiation patterns, and the peak gain and maximum efficiency over the operating band are 3.4 dBi and 95%, respectively.

**Keywords** Broadband, Compact, Coplanar Waveguide, C-band, Tapered Monopole Antenna

## 1. Introduction

In the recent years, with the rapid development of modern wireless and mobile communications, there has been great interest in wideband antenna designs. Many applications require antennas with compact size, multiband and/or broadband impedance bandwidth, stable gain and radiation pattern[1]. It is a well-known fact that microstrip planar antennas are characterized by their interesting features; lightweight, low-profile, low cost, small size and ease of fabrication. However, the conventional patch antenna suffers from a very narrow impedance bandwidth (less than 5%), which cannot satisfy the bandwidth requirement of the modern wireless systems[2]. Thus, different techniques for bandwidth enhancement are reported in literatures, such as the aperture coupled feed[3]–[5], stacked patches[6]–[9], coplanar coupled feed[10], L-probe feed[11],[12], E-shaped patch[13]–[16] and U-slot patch[9],[17]. By using these techniques, great enhancement on impedance bandwidth up to 50% has been attained for  $VSWR < 2$ . All these antennas based on slot structures are smaller than conventional monopole antennas but their structures are much more complex for practical applications.

Recently, many researchers have paid great attention for

improving the antenna bandwidth by employing the coplanar waveguide (CPW)-fed line since it has many features, such as simple structure, compact size, low cost, omnidirectional radiation pattern across the all operating bands and easy integration of monolithic microwave integrated circuits (MMIC)[18]. Hence, the CPW-fed planar slot antennas are identified as the most promising antenna design for wideband wireless applications[19]–[26]. Thus, CPW-fed method is adopted here to design the proposed antenna

In this paper, a new broadband antenna for C-band application is presented. The proposed antenna composes of a beveled CPW ground plane, taper-shaped monopole radiator and open-ended L-shaped slots. The L-shaped slots embedded in the monopole with tapered-structure are a new approach to broaden the bandwidth of an antenna. The principle and simulation analysis of the tapers and slots in achieving broadband bandwidth of an antenna is given in detail. Also, parametric study of the key parameters including the length and width of the slots and the indentation angles of the taper structures are also simulated and discussed. The broadband CPW-fed tapered monopole antenna (TMA) has a compact size of  $20 \times 18 \times 1.6 \text{ mm}^3$  and the achieved bandwidth covers 3.8-8 GHz for C-band application.

## 2. Proposed Antenna Configuration and Design Consideration

Normally, the bandwidth of the microstrip antenna is narrow since it has only one resonance frequency. The

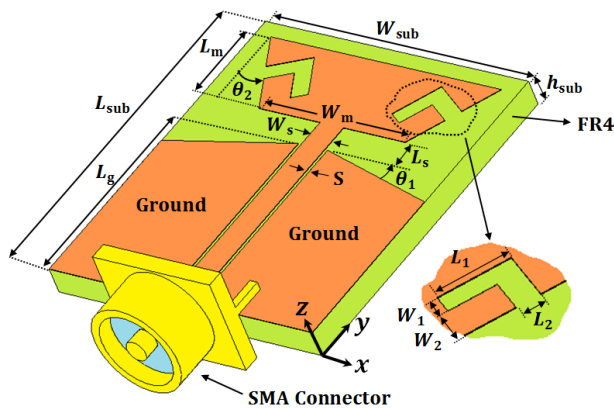
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bandwidth may be broadening if there more resonant parts are available with each one operating at its own resonant mode. Therefore, this leads to multiband or broadband performance due to the overlapping of these multiple resonances. Thus, a modification must be done in the geometry of a microstrip antenna to support the multiresonance. The principle that aforementioned above is adopted here for designing the proposed antenna. The configuration with detailed dimensions of the proposed CPW-fed tapered monopole antenna (TMA) is demonstrated in Fig. 1 and the various antenna structures involved in the design evolution process for broadband operation are shown in Fig. 2. All designed antennas are printed on a FR4 substrate of height  $h_{sub} = 1.6\text{mm}$  with relative permittivity,  $\epsilon_r = 4.3$  and the overall occupied area is characterized by substrate length and width,  $L_{sub}$  and  $W_{sub}$ , respectively, i.e.,  $(L_{sub} \times W_{sub})$  is  $20 \times 18\text{ mm}^2$ . The basis of the monopole radiator is a rectangular shape, which has the dimensions of length  $L_m = 5\text{mm}$  and width  $W_m = 10\text{ mm}$ . Afterwards, the structure is modified to form a tapered shape of indentation angle  $\theta_2 = 30^\circ$  at each side of the monopole width. The width of the CPW feed line is fixed at  $W_s = 1.6\text{ mm}$ , and a gap distance of  $S = 0.2\text{ mm}$  between the signal strip and the coplanar ground plane of length  $L_g = 12.25\text{mm}$  is used to achieve  $50\text{-}\Omega$  characteristic impedance. The feed line is terminated with a standard SMA connector. The TMA is connected to CPW-strip line and placed at a distance of  $L_s = 1.75\text{ mm}$  from the top of CPW ground plane. An open-ended L-slot, is positioned at the centre of tapered- monopole structures, and characterized by its lengths and widths,  $L_1 = 4.7\text{ mm}$ ,  $L_2 = 1.0\text{ mm}$  and  $W_1 = 1.0\text{ mm}$ ,  $W_2 = 1.7\text{ mm}$ , respectively. Finally, a pair of  $\theta_1 = 13.5^\circ$ -tapered structure is cut from the top of two-sided CPW ground plane. The electromagnetic simulation software CST Microwave Studio ver. 2012 is used to design the proposed antenna.

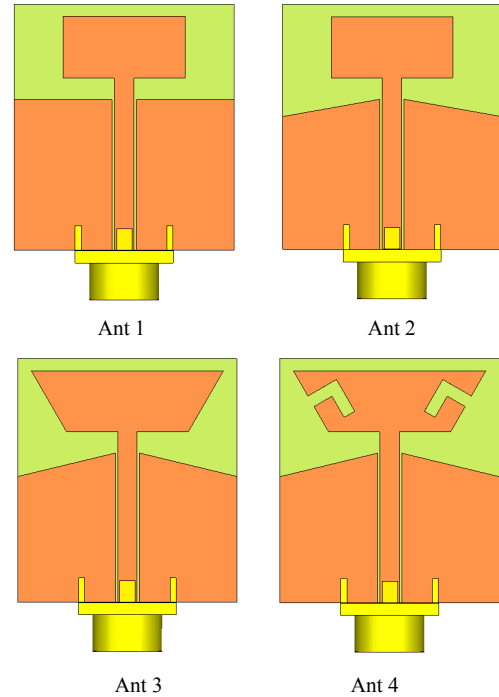


**Figure 1.** Geometry of the proposed CPW-fed slot antenna for C-band applications. ( $L_{sub} = 20$ ,  $W_{sub} = 18$ ,  $L_g = 12.25$ ,  $L_m = 5$ ,  $W_m = 10$ ,  $L_1 = 4.7$ ,  $W_1 = 1.0$ ,  $L_2 = 1.0$ ,  $W_2 = 1.7$ ,  $L_s = 1.75$ ,  $W_s = 1.6$ ,  $\theta_1 = 13.5^\circ$ ,  $\theta_2 = 30^\circ$ ,  $S = 0.2$ ,  $h_{sub} = 1.6$ ). (Unit: mm)

Antenna 4 (proposed antenna) is designed after Ant 1, 2 and 3 are designed, firstly. Antenna 1 in Fig. 2 represents the reference antenna that initially designed by assuming that the thickness of the substrate  $h_{sub}$  and its dielectric constant  $\epsilon_r$ , and the lowest frequency of operation ( $f_l = 4\text{ GHz}$ ) are given.

Then, substrate length ( $L_{sub}$ ) and substrate width ( $W_{sub}$ ) of the antenna are calculated as

$$L_{sub} = W_{sub} = \frac{c}{2f_l \sqrt{\frac{\epsilon_r + 1}{2}}} \quad (1)$$



**Figure 2.** Four improved prototypes of the proposed CPW-fed slot antenna

where  $c$  is speed of light. Note that the length and width of the antenna structure, according to (1), are equal to half of the effective wavelength at the lower frequency  $f_l$ . From (1), one can found that  $L_{sub} = W_{sub} = 23\text{ mm}$ . Afterwards, the CST software is used to sweep each geometric parameter and then the results are fine-tuned until the required resonance frequency with reasonable bandwidth is achieved. As shown in Fig. 3, this design can excite one resonant mode near  $5.2\text{GHz}$  with  $-10\text{dB}$  bandwidth of  $2\text{ GHz}$  ranging from  $4.3$  to  $6.3\text{GHz}$ . In order to broaden the antenna bandwidth to satisfy C-band operation ( $4\text{-}8\text{ GHz}$ ), Ant 2 is realized by cutting the CPW ground of Ant 1 by tapered structure of angle  $\theta_1 = 13.5^\circ$ . It generates two-resonant modes, one at  $4.85\text{ GHz}$  and the other at  $5.8\text{ GHz}$  and is capable of operating over the  $4.2\text{--}6.8\text{ GHz}$  frequency range, as noticed in Fig. 3. Thus, the bandwidth is increased without enlarging the size of antenna. It is seen that the lower frequency range  $f_l = 4.2\text{ GHz}$  is nearly unaffected, whereas, the upper frequency range  $f_h$  is increased from  $6.3\text{ GHz}$  to  $6.8\text{ GHz}$  when Ant 2 is introduced. It needs other modification in Ant 2 to increase  $f_h$  toward  $8\text{ GHz}$ , the upper range of C-band application. Therefore, Ant 3 is proposed for this purpose, which contains other tapered structure of angle  $\theta_2 = 30^\circ$  added to the vertical side of the monopole antenna. As seen from Fig. 3, two resonant-modes of  $4.3$  and  $6.0\text{ GHz}$  are produced and more bandwidth of  $3.1\text{ GHz}$  ranging from  $3.9$  to  $7.0\text{ GHz}$  is achieved compared with Ant 2 bandwidth of  $2\text{ GHz}$ . It is still

that higher frequency range has not satisfied; this leads to introduce Ant 4 to perform this requirement. Open-ended L-slots are added to the tapered-monopole structure of Ant3 to increase the perimeter length of the radiating structure which enhances the antenna bandwidth significantly. This can be seen from Fig. 3, where the bandwidth of C-band application is performed which is about of 4.2 GHz operates

from 3.8 to 8.0 GHz. The reason behind that is due to the generation of higher resonant-mode of 7.8 GHz in addition to the lower resonant-modes of 4.2 and 6.3 GHz that achieved from Ant 3. Thus, the Ant 4 (proposed antenna) is capable of satisfying the C-band application, i.e., (4 – 8 GHz) band.

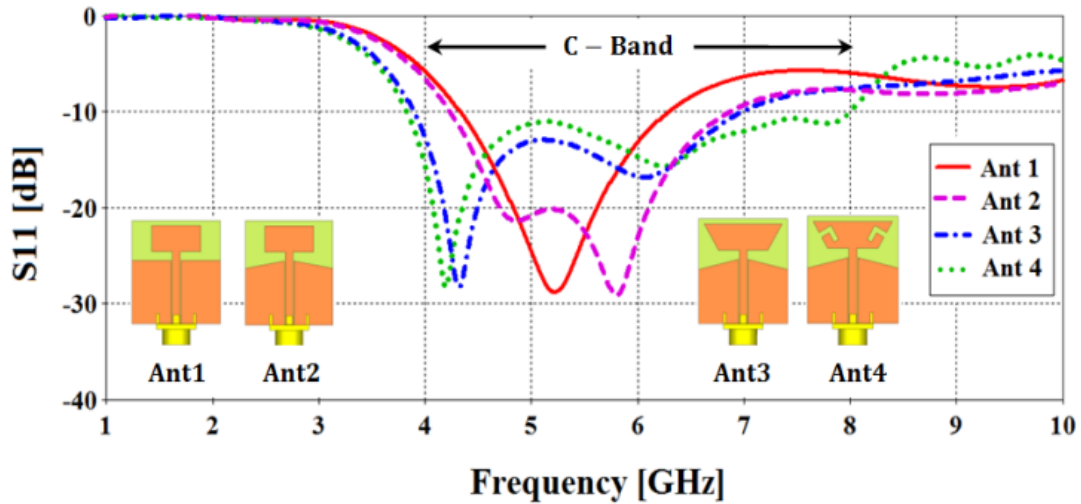
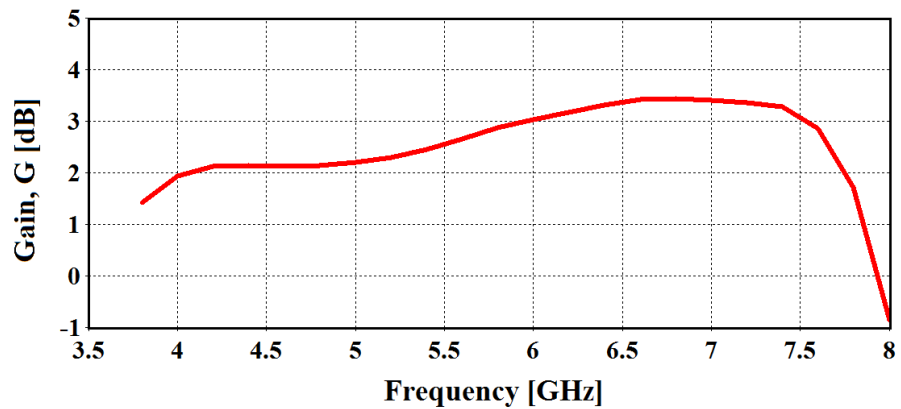
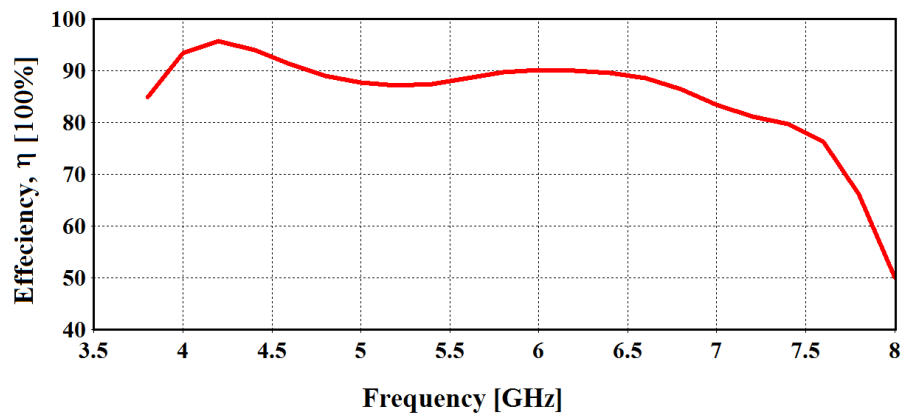


Figure 3. Simulated  $S_{11}$  results for antennas 1-4 (for CPW-fed slot antenna with optimized values of the designed parameters)



(a)



(b)

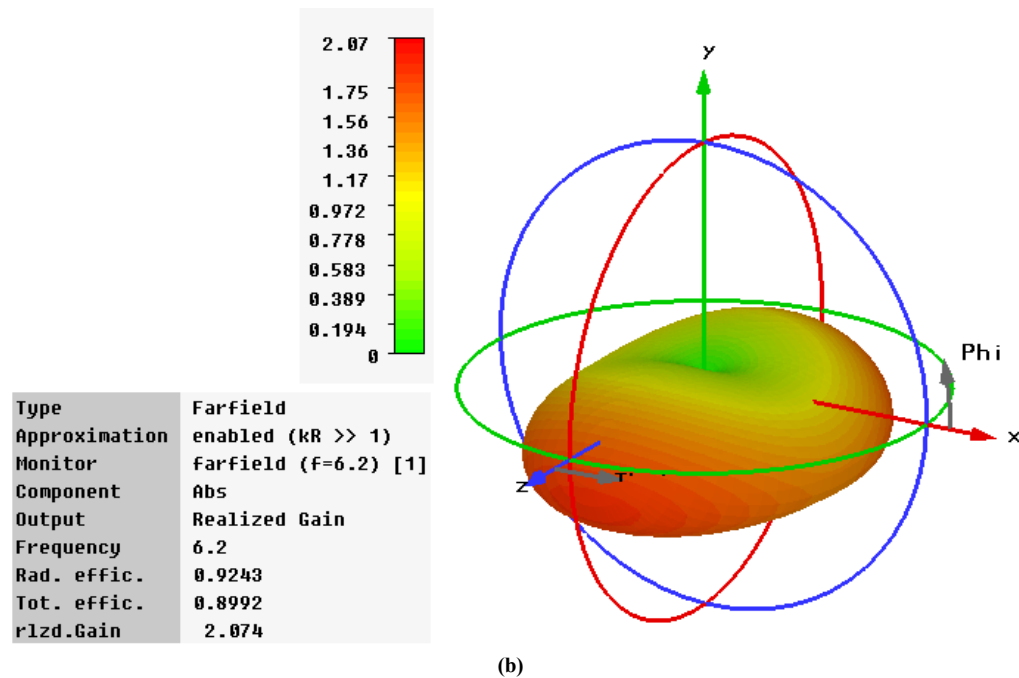
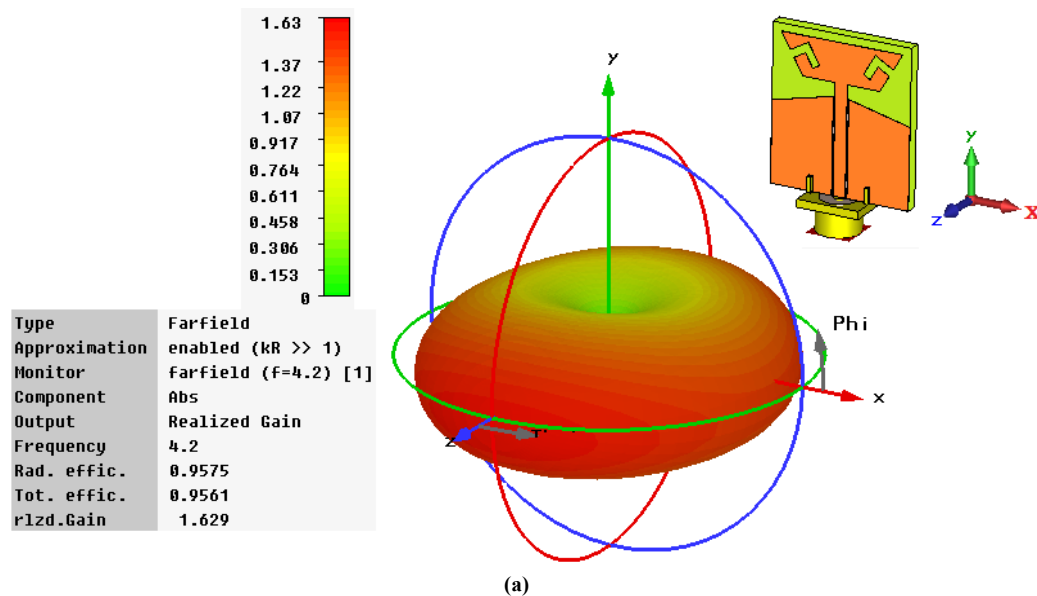
Figure 4. Simulated results of (a) gain (G) and (b) efficiency ( $\eta$ ) of the proposed antenna (Antenna 4)

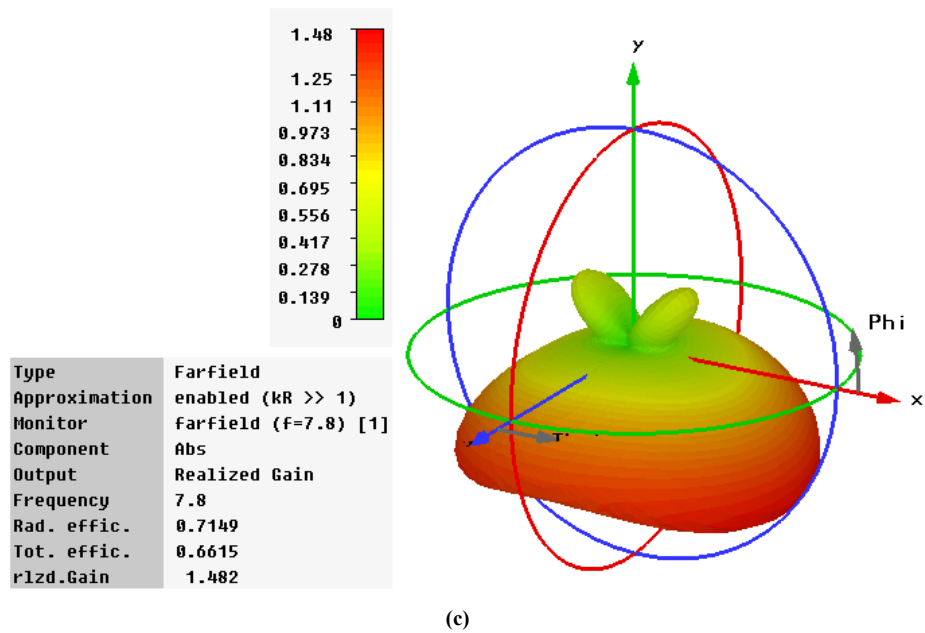
### 3. Simulated Results and Discussion

The radiation characteristics such as antenna peak gain and radiation patterns across the operating bands for the proposed antenna have also been presented and discussed. The simulated gain and efficiency variations with frequency are shown in Fig. 4(a) and (b), respectively. The obtained gain and efficiency are between -0.8 and 3.4 dBi, and 50 and 95%, respectively with nearly stable gain are obtained about 2 dBi and 3 dBi over frequency ranges, 4 – 5 GHz and 6 – 7.5 GHz, respectively. The simulated radiation patterns are shown in Figs. 5 and 6. The 3D and 2D patterns are simulated and depicted at three different frequencies (resonant-modes of Ant 4): 4.2, 6.2, and 7.8 GHz. Fig. 5

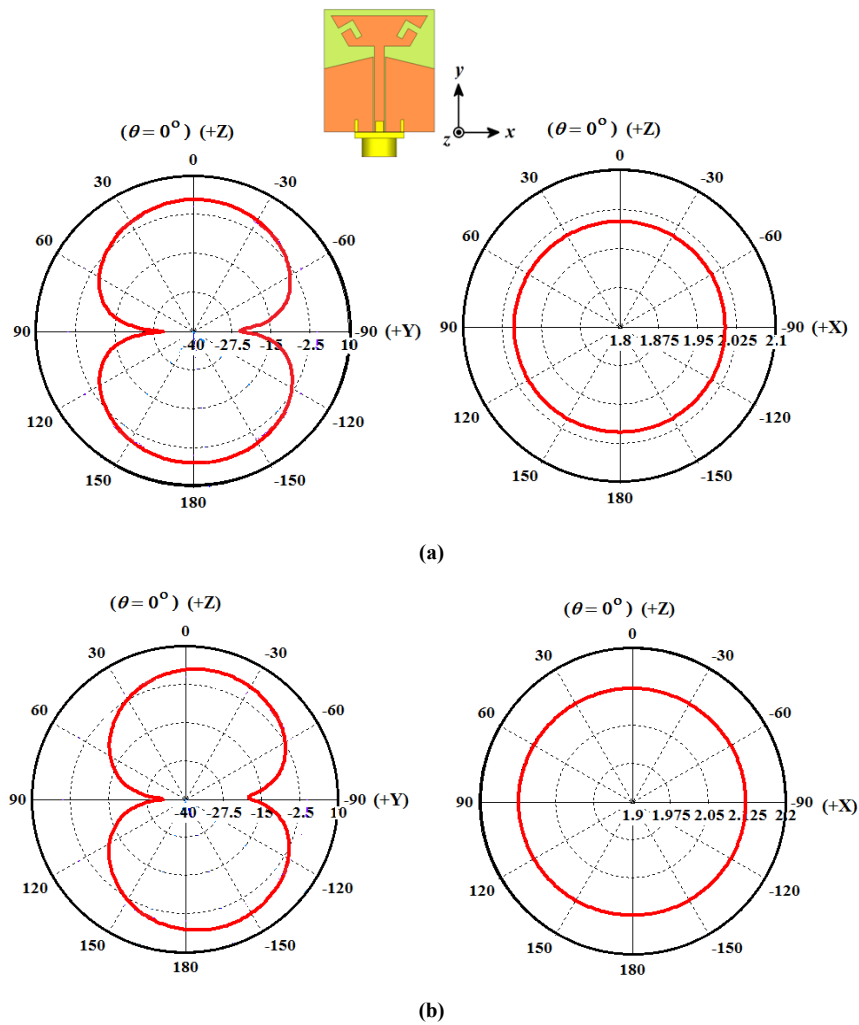
presents 3D farfield gain patterns for the aforementioned three frequencies, as shown, an omnidirectional and stable radiation patterns are achieved for low and middle frequency bands, see Fig. 5(a) and (b). But at high frequency band, Fig. 5(c), nearly omnidirectional with unsymmetrical pattern about xz-plane is obtained. The 2D absolute gain patterns in the yz- or E-plane and xz- or H-plane are seen in Fig. 6 for frequencies, 4.2, 6.2 and 7.80.

GHz. It seen that an omnidirectional pattern are sustained in H-plane (xz-plane) over all the frequencies in the C-band. But a figure of 8 is obtained for lower and middle frequencies, Fig. 6(a) and (b), and distorted version of figure 8 is seen for higher frequency in the band, see Fig. 6(c).





**Figure 5.** Simulated results of 3D farfield gain patterns at (a)  $f_r = 4.2$  GHz (b)  $f_r = 6.2$  GHz and (c)  $f_r = 7.8$  GHz for the proposed antenna (Antenna 4)



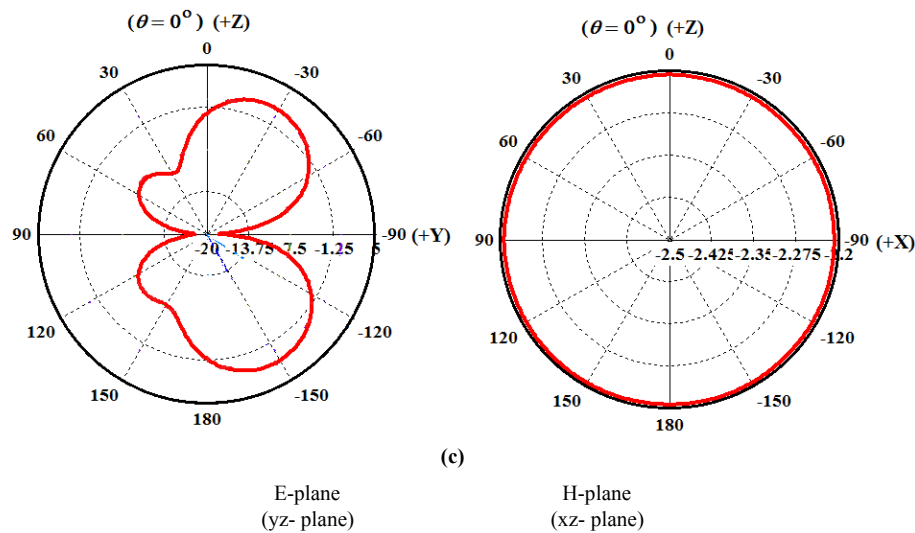


Figure 6. Simulated radiation patterns for the proposed antenna (Ant 4) at (a) 4.2 GHz (b) 6.2 GHz (c) 7.8 GHz

## 4. Key Parameters Study

In this section, the key parameters of the antenna are investigated to show their influences on the antenna performance. The angles  $\theta_1$  and  $\theta_2$  of the tapered monopole and CPW ground, respectively, and the lengths and widths of L-shaped slot  $L_1$  and  $L_2$ ,  $W_1$  and  $W_2$ , respectively. These parameters are especially examined to study their influences on impedance bandwidth. All other parameters keep their initial values as seen in Fig. 1.

### 4.1. Effects of the Tapered Structures

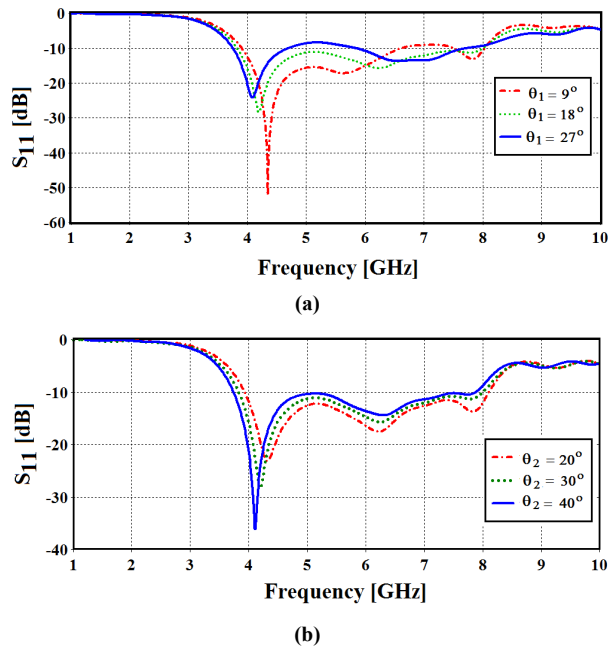


Figure 7. The simulation results for  $S_{11}$  as affected by parameter (a)  $\theta_1$  (b)  $\theta_2$

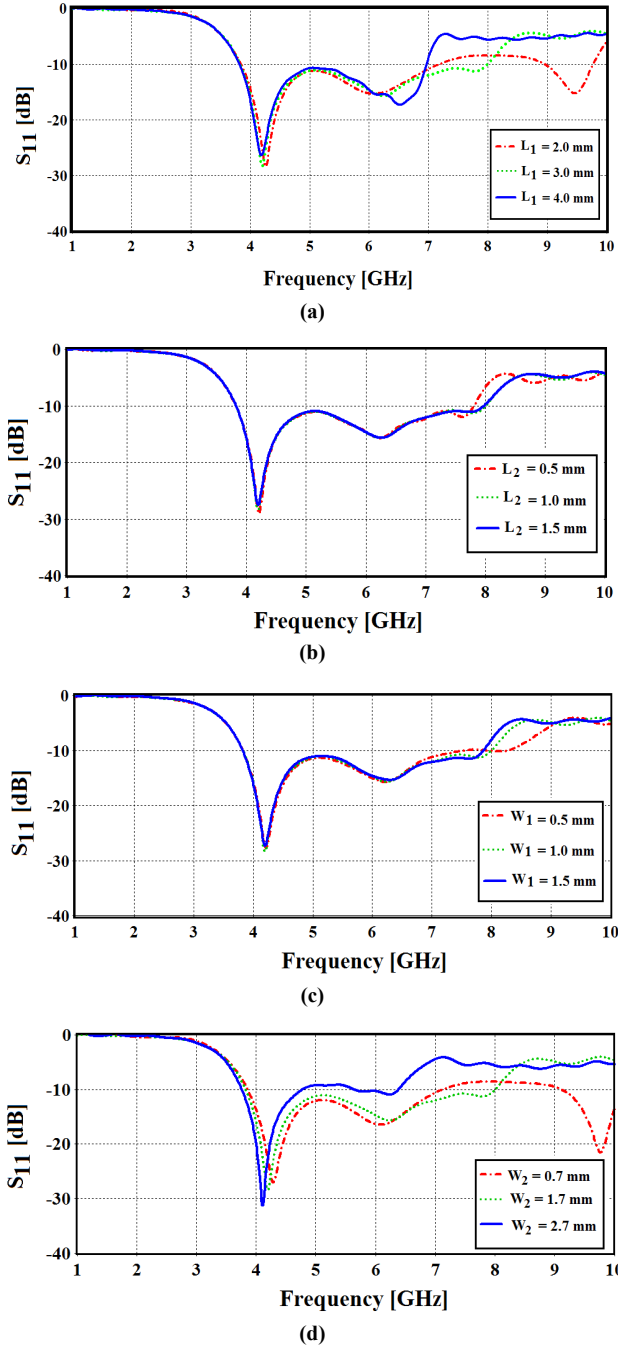
In Fig. 7, the two parameters, angles  $\theta_1$  and  $\theta_2$ , of the

taper-shaped structures are investigated to find their effects on the antenna performance. We can see that the antenna's return loss is greatly affected by the ground CPW taper angle  $\theta_1$ , Fig. 7(a), while less effect is observed for the tapered monopole angle  $\theta_2$  on antenna's return loss, Fig. 7(b). As  $\theta_1$  is increased from  $9^\circ$  to  $27^\circ$ , the lower resonant-mode is increased from 4.1 to 4.3 GHz with best value  $\theta_1 = 18^\circ$  for 4.2 GHz resonant-mode, but higher resonant-mode 7.8 GHz is unaffected for angles less than  $18^\circ$  and it disappears for angles greater than  $18^\circ$ . In Fig. 7(b), lower resonant-mode is decreased when  $\theta_2$  is increases while higher resonant-mode is decreased slightly with increasing in  $\theta_2$ .

### 4.2. Effects of the L-Shaped Slot

The effect of the important parameter, L-shaped slot, on the antenna performance, return loss is demonstrated in Figs. 8(a)–(b). It is seen from Fig. 8(a), the length  $L_1$  influences only on higher resonant-mode and not affect on lower resonant-mode. For  $L_1 = 2$  mm and 4 mm, the higher frequency band is reduced from 8 GHz, for  $L_1 = 3$  mm, to 7 GHz and the bandwidth is only 3 GHz. The length  $L_2$  has nearly not changes the return loss when it is varying from 0.5 mm to 1.5 mm, see Fig. 8(b). Also, as seen from Fig. 8(c), the width  $W_1$  is approximately has not affect on the antenna's return loss when it is varying from 0.5 mm to 1.5 mm. Finally, as depicted in Fig. 8(d), the width  $W_2$  has little effect on lower resonant-mode, but more effect has been observed at higher resonant-mode, hence, on the bandwidth when it is varying from 0.7–2.7 mm. It is concluded from the previous discussion that the tapers and L-shaped structures have great influences on the higher-resonant mode and less affect on the lower resonance-mode. Also, three out of the six parameters,  $\theta_1$ ,  $L_1$  and  $W_2$  play an important rule for achieving great enhancement in antenna bandwidth.





**Figure 8.** The simulation results for  $S_{11}$  as affected by parameter (a)  $L_1$  (b)  $L_2$  (c)  $W_1$  (d)  $W_2$

## 5. Conclusions

In this paper, coplanar waveguide (CPW)-fed tapered monopole antenna (TMA) for C-band application (4 – 8 GHz) has been proposed and simulated using taper and L-shaped structure. The concept of resonant-mode frequencies overlapping technique is used to get wide bandwidth via adding resonant geometrical parts to the antenna. A compact size of TMA structure is obtained from the design process ( $20 \times 18 \text{ mm}^2$ ). Additionally, the antenna exhibits a simulated dB return-loss bandwidth of 4.2

GHz ranging from 3.8 to 8 GHz, while maintaining a high efficiency and gain in the order of 95% and 3.4 dBi at the bands of interest. It is therefore well suited for C-band applications. These applications include IEEE 802.11b/g wireless network standards (5.15-5.35 GHz and 5.725-5.825 GHz), communications satellites, satellite radio, weather radar, etc.

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